

杭州云镓半导体科技有限公司
Hangzhou CloudSemi Technology Co., Ltd

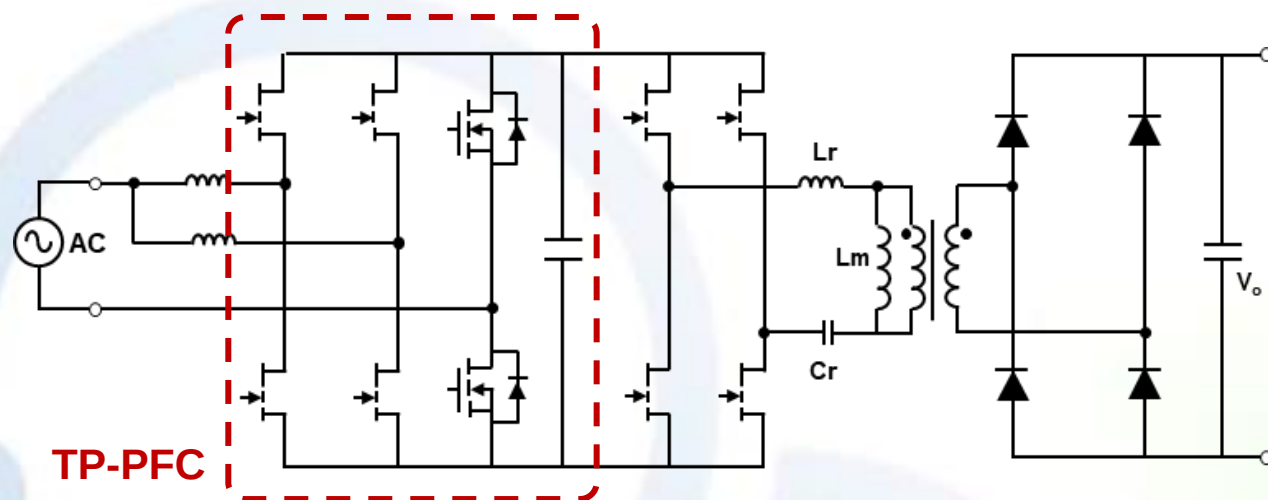
GaN FET Loss Calculation in Boost Converter

GaN Device R&D Team

Powering the Dreams!



Zero Q_{rr} enables GaN to work in CCM TP-PFC



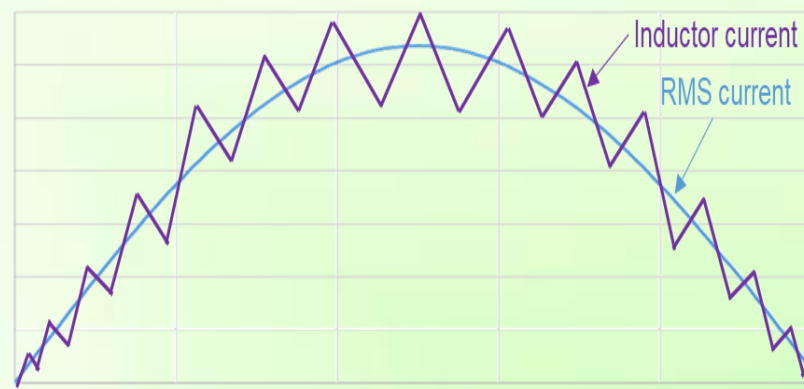
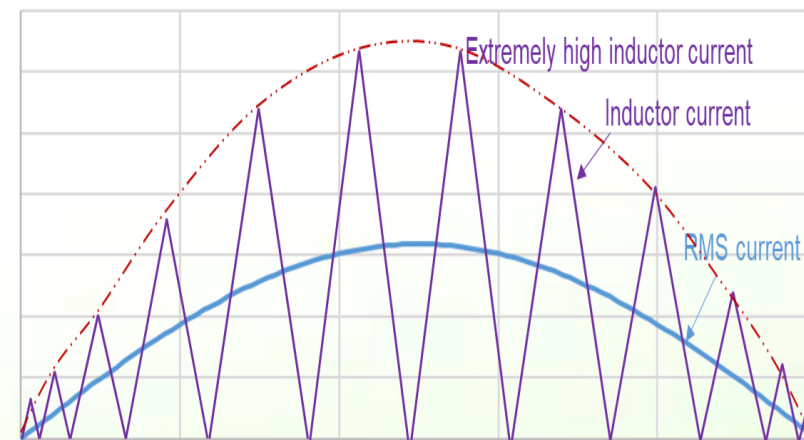
Enhancement-mode GaN FET:

Low Q_{oss} → Low Q_{oss} loss

Low Q_{oss} → High switching frequency

Zero Q_{rr} → No Q_{rr} loss → high efficiency

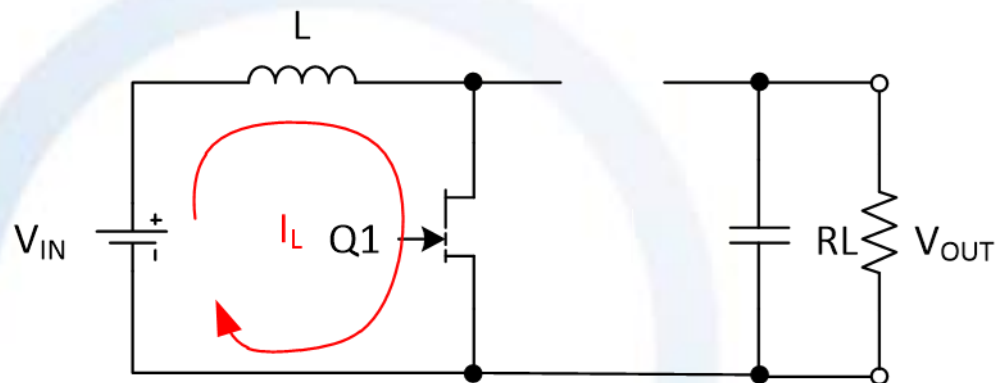
Zero Q_{rr} enables GaN work in CCM mode, then help reduce system cost



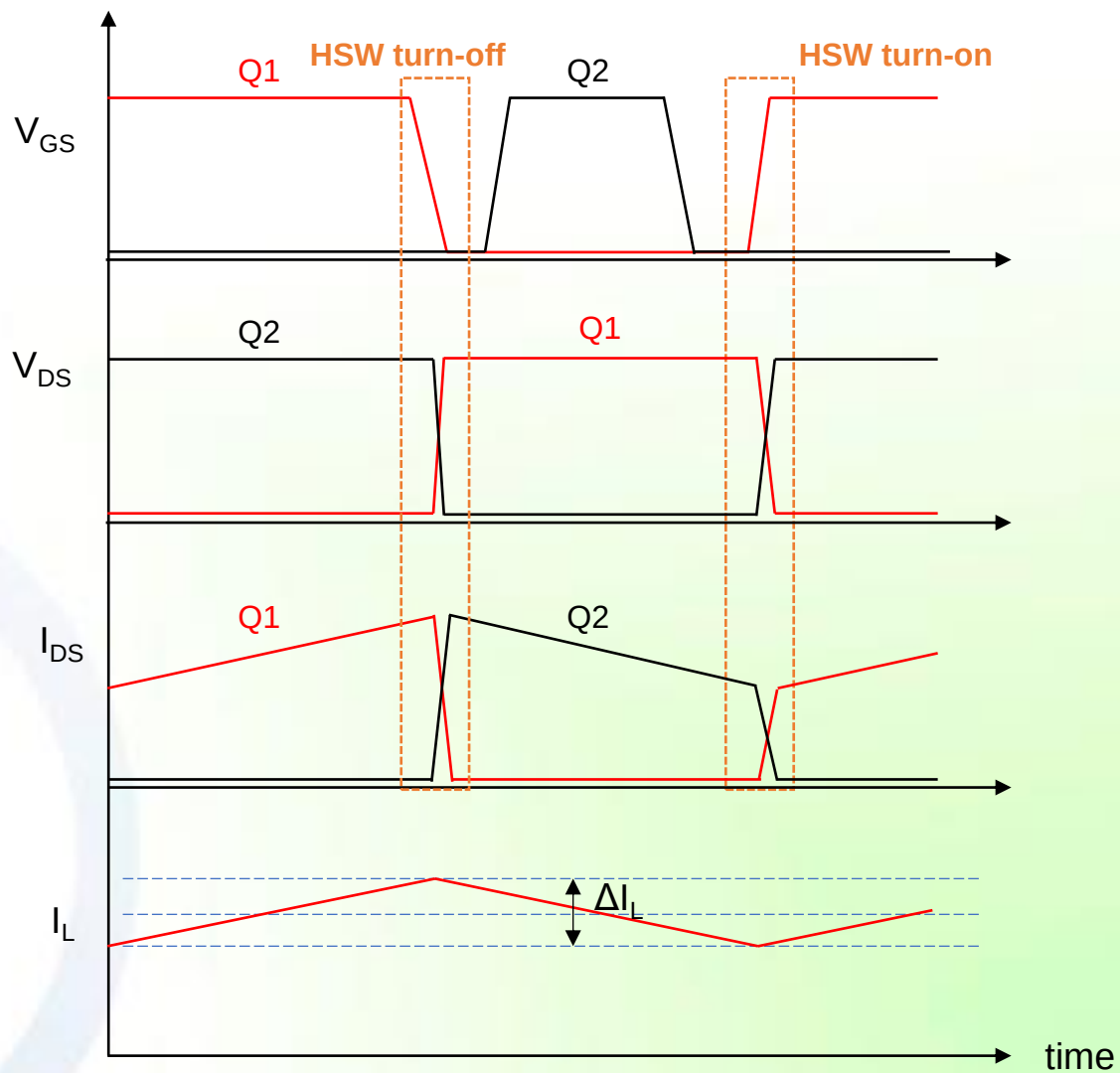
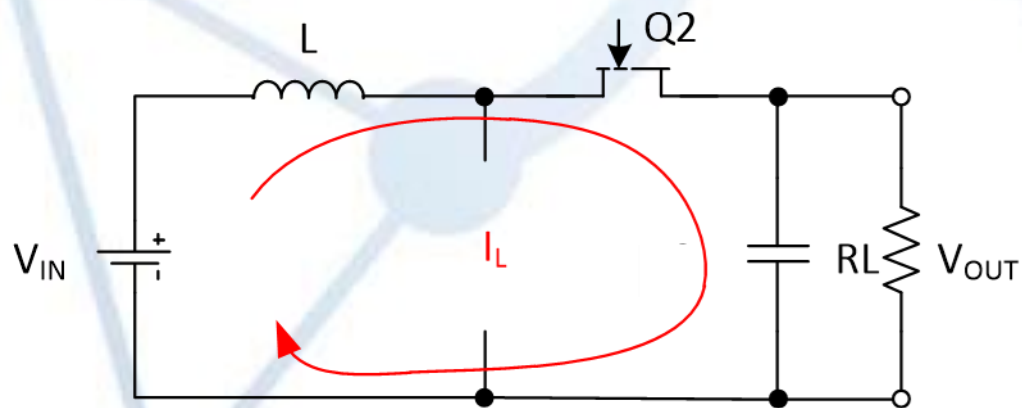
In this slide, GaN FET loss is analyzed using a synchronous boost converter (CCM mode)

Synchronous CCM boost converter – waveforms

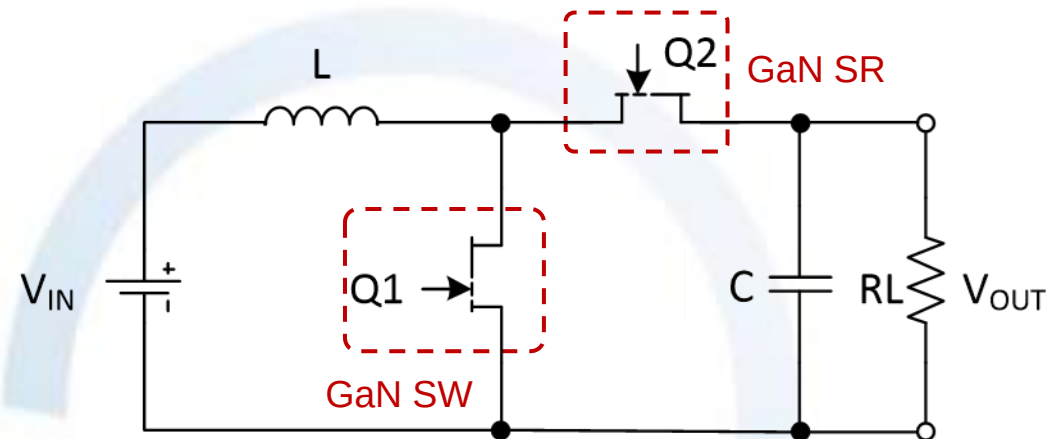
When Q1 turns on



When Q1 turns off



Synchronous CCM boost converter – key equations



When Q1 turns on:

$$\frac{dI_L}{dt} = \frac{V_{IN}}{L}$$

When Q1 turns off:

$$\frac{dI_L}{dt} = \frac{V_{OUT} - V_{IN}}{L}$$

According to *Volt-Second Balance*:

$$V_{OUT} = V_{IN} * \frac{1}{1 - D}$$

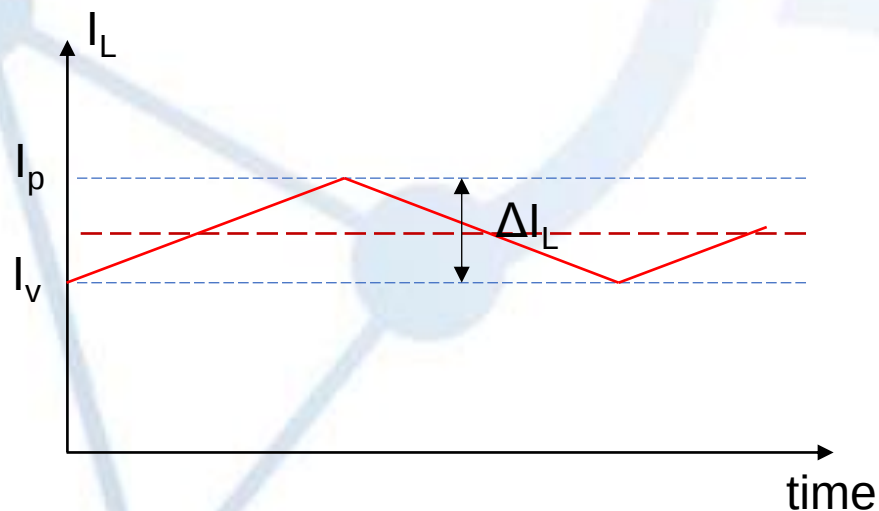
Output current vs inductor current

$$I_O = I_L * (1 - D)$$

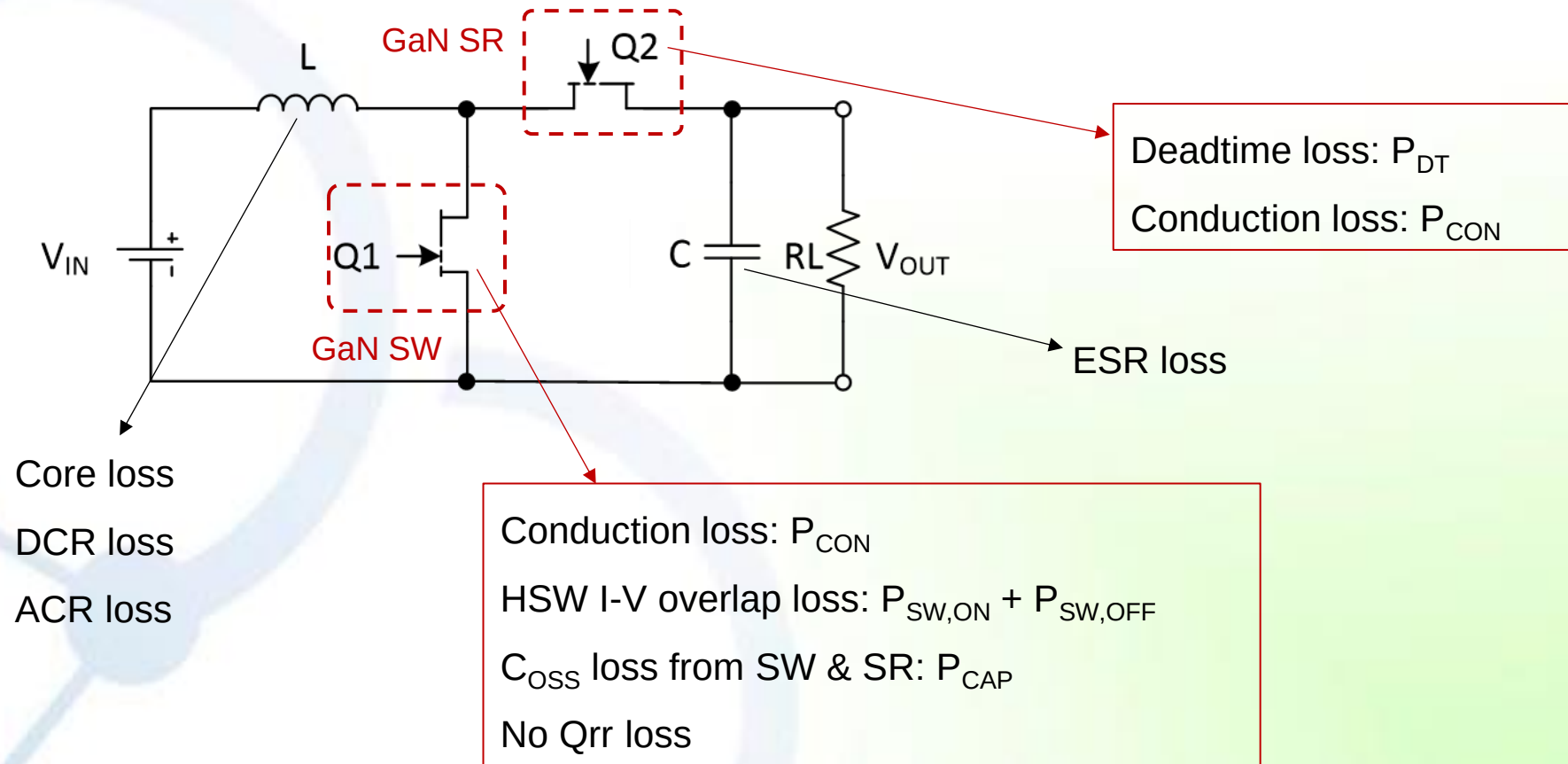
Ripple current through inductor ΔI_L

$$\Delta I_L = \frac{V_{IN}}{L} * DT = I_p - I_v$$

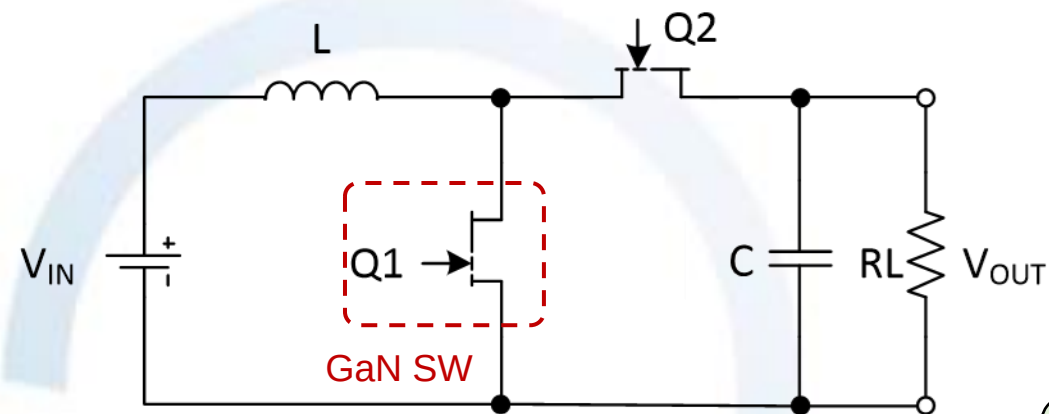
$$I_L = \frac{1}{2} * (I_p + I_v)$$



Loss breakdown of GaN FETs



Loss breakdown of GaN FETs: SW conduction loss

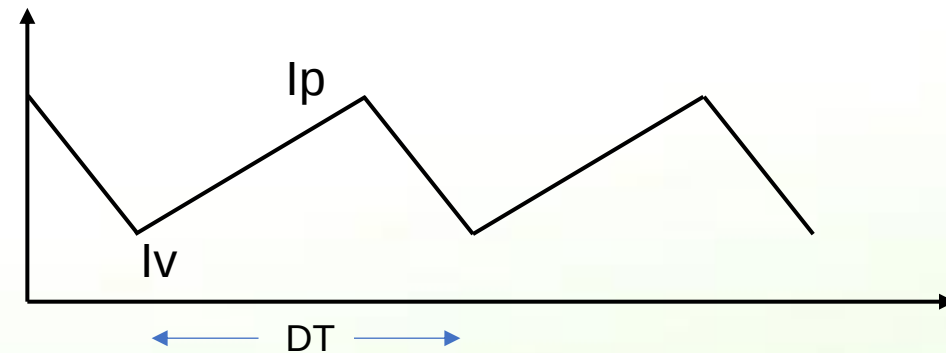


$$I_p = I_L + \frac{\Delta I_L}{2} \quad I_v = I_L - \frac{\Delta I_L}{2}$$

$$I_{rms, sw} = \sqrt{\frac{I_p^2 + I_v^2 + I_p I_v}{3}} * D$$

$$P_{cond, sw} = I_{rms, sw}^2 * R_{dson} = \frac{I_p^2 + I_v^2 + I_p * I_v}{3} * D * R_{dson}$$

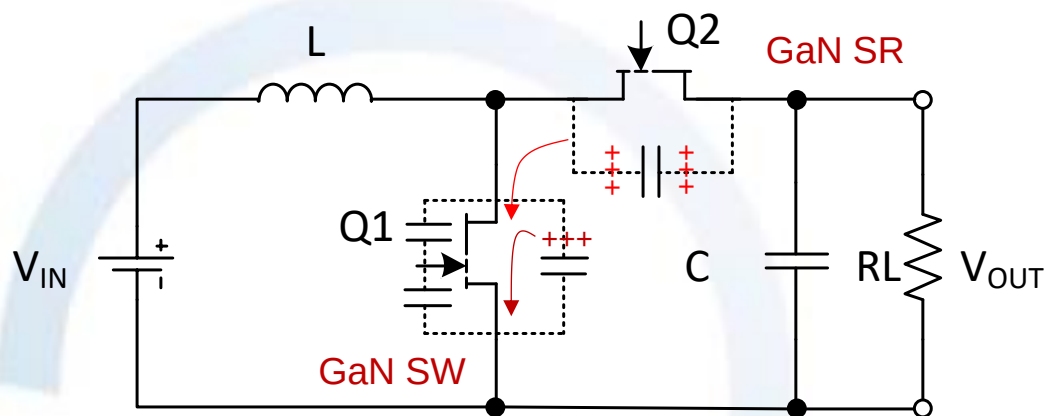
RMS current derivation



$$\begin{aligned} \int_0^{DT} i(t)^2 dt &= \int_0^{DT} \left(I_v + \frac{I_p - I_v}{DT} t \right)^2 dt \\ &= \int_0^{DT} \left(I_v^2 + \frac{(I_p - I_v)^2}{(DT)^2} t^2 + 2 \frac{I_v I_p - I_v^2}{DT} t \right) dt \\ &= \left(I_v^2 t + \frac{(I_p - I_v)^2}{3(DT)^2} t^3 + \frac{I_v I_p - I_v^2}{DT} t^2 \right) \Big|_0^{DT} \\ &= I_v^2 DT + \frac{I_p^2 + I_v^2 - 2I_p I_v}{3} DT + (I_v I_p - I_v^2) DT \\ &= \frac{I_p^2 + I_v^2 + I_p I_v}{3} DT \end{aligned}$$

$$I_{rms} = \sqrt{\frac{I_p^2 + I_v^2 + I_p I_v}{3}} D$$

Loss breakdown of GaN FETs: SW+SR C_{OSS} loss



During SW turn-on transient,

C_{OSS} of SW-Q1 is discharged through SW-Q1's channel (Current CANNOT be captured)

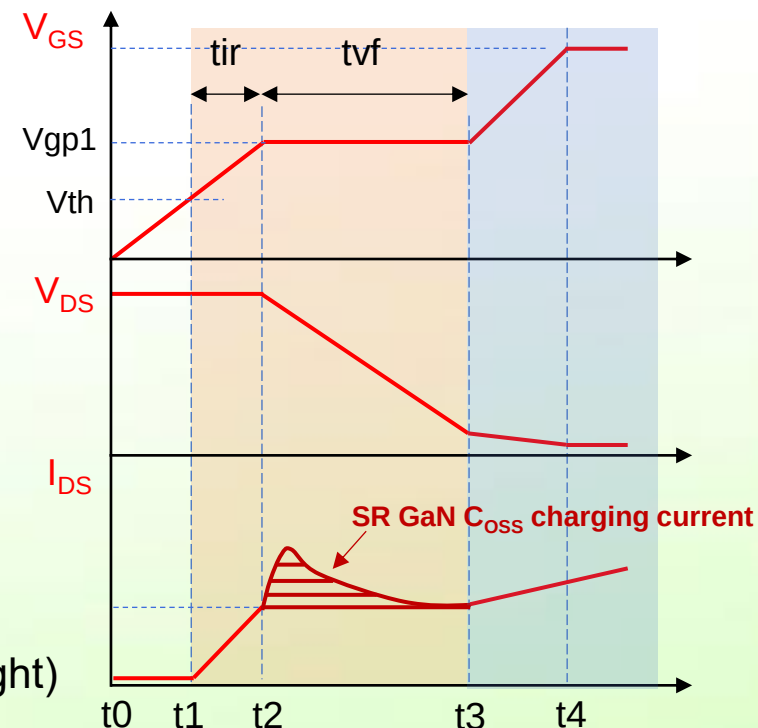
C_{OSS} of SR-Q2 is charged through SW-Q1's channel (Current CAN be captured, as in right)

Both two capacitors charge behavior **generates heat on SW-Q1**.

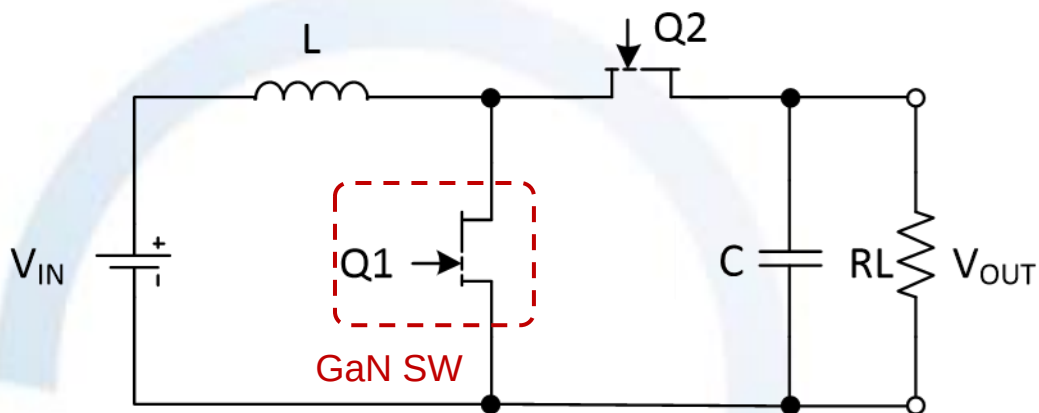
$$Cap\ loss(SW) = \int_0^{V_{bus}} C_{OSS} * V_{ds} dV \quad Cap\ loss(SR) = \int_0^{V_{bus}} C_{OSS} * (V_{bus} - V_{ds}) dV = V_{bus} * Q_{oss} - \int_0^{V_{bus}} C_{OSS} * V_{ds} dV$$

If Q1 and Q2 features the same part, then the C_{OSS} loss on SW-Q1 will be:

$$P_{cap} = (Cap_loss_SW + Cap_loss_SR) * f_s = V_{bus} * Q_{OSS} * f_s$$



Loss breakdown of GaN FETs: SW I-V overlap loss



Roughly based on right schematic waveforms:

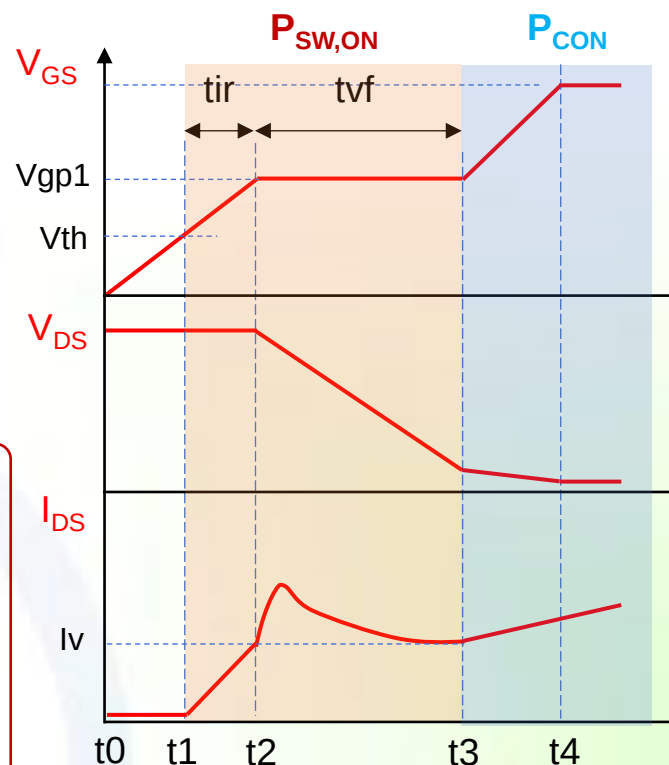
$$P_{SW,ON} \approx \frac{1}{2} * V_{IN} * I_v * t_{on} * f_s$$

$$P_{SW,OFF} \approx \frac{1}{2} * V_{IN} * I_p * t_{off} * f_s$$

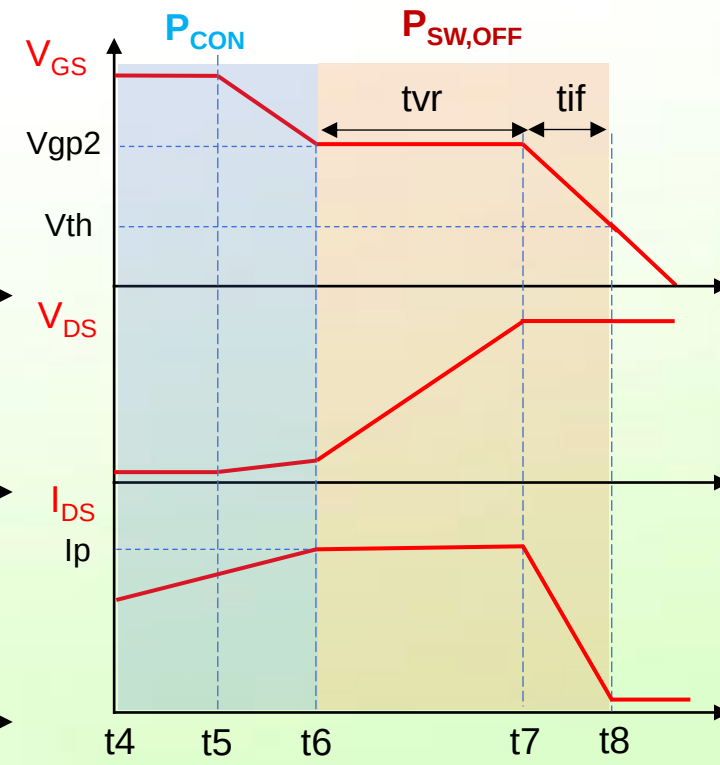
$$P_{SW} \approx \frac{1}{2} * V_{IN} * I_v * t_{on} * f_s + \frac{1}{2} * V_{IN} * I_p * t_{off} * f_s$$

$$t_{on} \approx t_{ir} + t_{vf} \quad t_{off} \approx t_{vr} + t_{if}$$

Turn-on transient



Turn-off transient



Turn-on and turn-off time (t_{on} , t_{off}) also can be calculated from capacitance charging & discharging analysis.

Switching time calculation - tir

During t0~t2

According to equivalent circuit

$$I_g = I_{gs} + I_{gd} = \frac{V_{GS} - V_{gs}}{R_g}$$

$$I_{gs} = C_{gs} \frac{dV_{gs}}{dt}$$

$$I_{gd} = C_{gd} \frac{d(V_{gs} - V_{DS})}{dt} = C_{gd} \frac{d(V_{gs})}{dt}$$

Then

$$I_g = C_{gs} \frac{dV_{gs}}{dt} + C_{gd} \frac{d(V_{gs})}{dt} = (C_{gs} + C_{gd}) \frac{d(V_{gs})}{dt}$$

$$\frac{d(V_{gs})}{V_{GS} - V_{gs}} = \frac{dt}{R_g(C_{gs} + C_{gd})}$$

$$-\ln(V_{GS} - V_{gs}) = \frac{t}{R_g(C_{gs} + C_{gd})} + k$$

$$V_{gs} = V_{GS} - k e^{\frac{-t}{R_g(C_{gs} + C_{gd})}}$$

So

$$V_{gs} = V_{GS} * \left(1 - e^{\frac{-t}{R_g(C_{gs} + C_{gd})}}\right)$$

$$e^{\frac{-t}{R_g C_{iss}}} = 1 - \frac{V_{gs}}{V_{GS}}$$

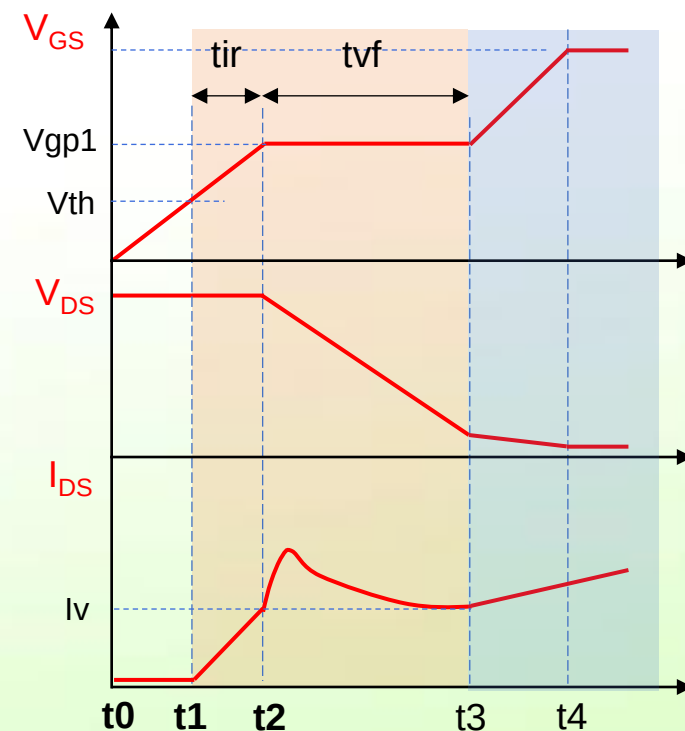
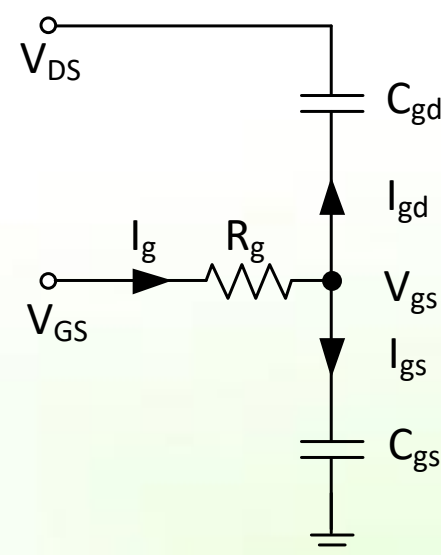
$$t = R_g C_{iss} * \ln\left(\frac{1}{1 - \frac{V_{gs}}{V_{GS}}}\right) = R_g C_{iss} * \ln\left(\frac{V_{GS}}{V_{GS} - V_{gs}}\right)$$

Finally (R_G includes R_g and external resistance R_{gon})

$$t_1 - t_0 = R_G C_{iss} * \ln\left(\frac{V_{GS}}{V_{GS} - V_{TH}}\right)$$

$$t_2 - t_0 = R_G C_{iss} * \ln\left(\frac{V_{GS}}{V_{GS} - V_{gp1}}\right)$$

equivalent circuit



$$t_{ir} = t_2 - t_1 = R_G C_{iss} * \ln\left(\frac{V_{GS} - V_{TH}}{V_{GS} - V_{gp1}}\right)$$

$$V_{gp1} = V_{TH} + \frac{I_v}{g_m}$$

Switching time calculation - tvf

During t2~t3

According to equivalent circuit

$$I_g = I_{gd}$$

$$I_{gd} = C_{gd} \frac{d(V_{ds})}{dt}$$

Then

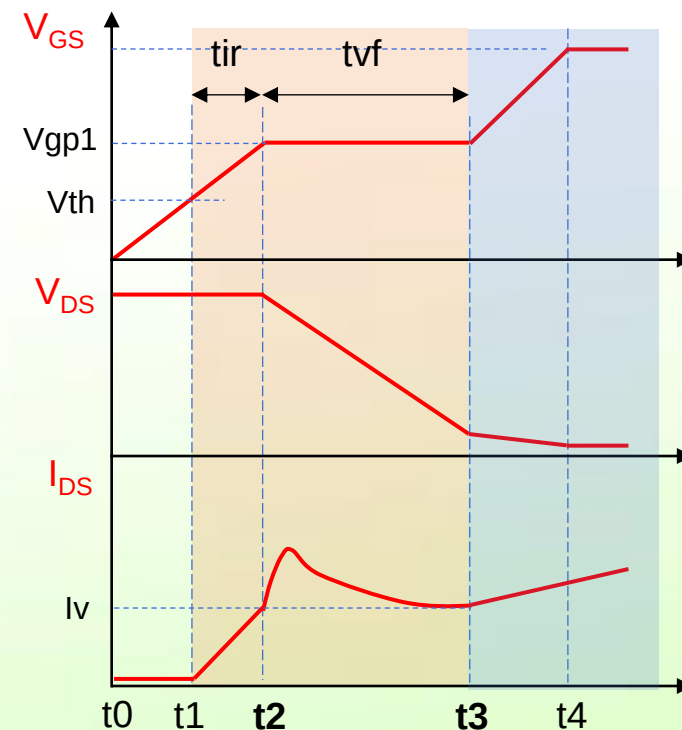
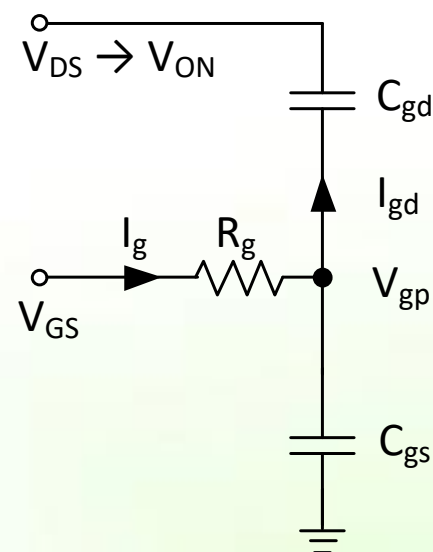
$$\frac{V_{GS} - V_{gs}}{R_g} = C_{gd} \frac{d(V_{ds})}{dt}$$

$$dt = R_g C_{gd} * \frac{d(V_{ds})}{(V_{GS} - V_{gs})}$$

Finally (R_G includes R_g and external resistance R_{gon})

$$t_3 - t_2 = R_G C_{gd} * \frac{V_{DS}}{V_{GS} - V_{gp1}}$$

equivalent circuit



$$t_{vf} = t_3 - t_2 = R_G C_{gd} * \frac{V_{DS}}{V_{GS} - V_{gp1}}$$

$$t_{vf} = R_G * \frac{Q_{GD}}{V_{DS}} * \frac{V_{DS}}{(V_{GS} - V_{gp1})}$$

$$V_{gp1} = V_{TH} + \frac{I_v}{g_m}$$

Switching time calculation – t_{vr} & t_{if}

Similar to the analysis of turn-on transient
(R_G includes R_g and external resistance R_{goff})

$$t_6 - t_5 = R_G C_{iss} * \ln\left(\frac{V_{GS}}{V_{gp2}}\right)$$

$$t_7 - t_6 = R_G C_{gd} \frac{V_{DS}}{V_{gp2}}$$

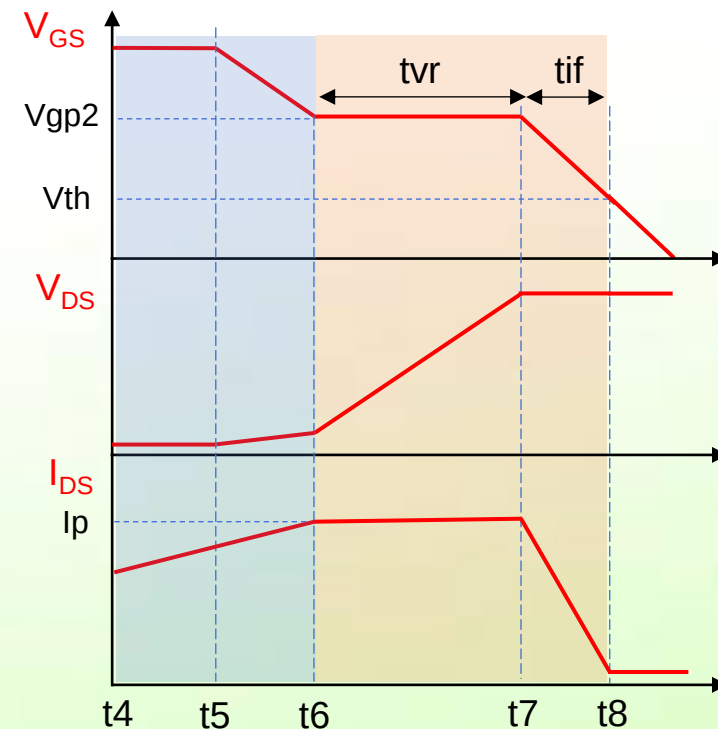
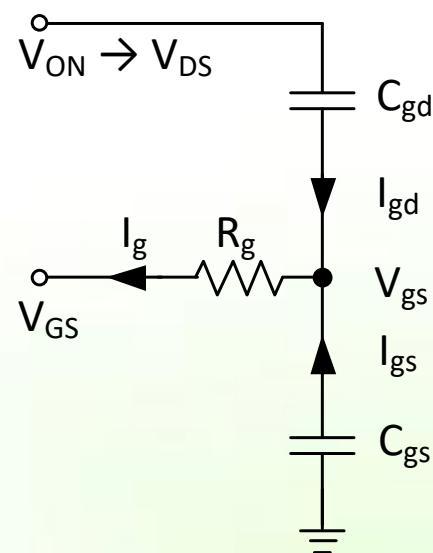
$$t_8 - t_7 = R_G C_{iss} \ln\left(\frac{V_{gp2}}{V_{TH}}\right)$$

$$t_{vr} = t_7 - t_6 = R_G * \frac{Q_{GD}}{V_{DS}} * \frac{V_{DS}}{V_{gp2}}$$

$$t_{if} = t_8 - t_7 = R_G * C_{iss} * \ln\left(\frac{V_{gp2}}{V_{TH}}\right)$$

$$V_{gp2} = V_{TH} + \frac{I_p}{g_m}$$

equivalent circuit



* Miller plateau V_{gp2} during turn-off transient is not the same as turn-on because of the different switching current.

Switching time summary

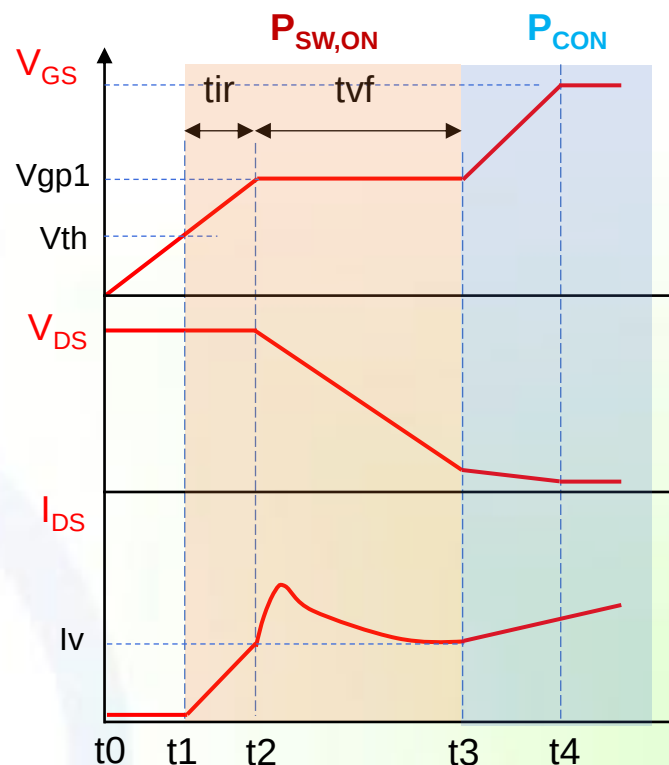
$$t_{ir} = R_G C_{iss} * \ln \left(\frac{V_{GS} - V_{TH}}{V_{GS} - V_{gp1}} \right)$$

$$t_{vf} = R_G \frac{Q_{GD}}{V_{DS}} * \frac{V_{DS}}{(V_{GS} - V_{gp1})}$$

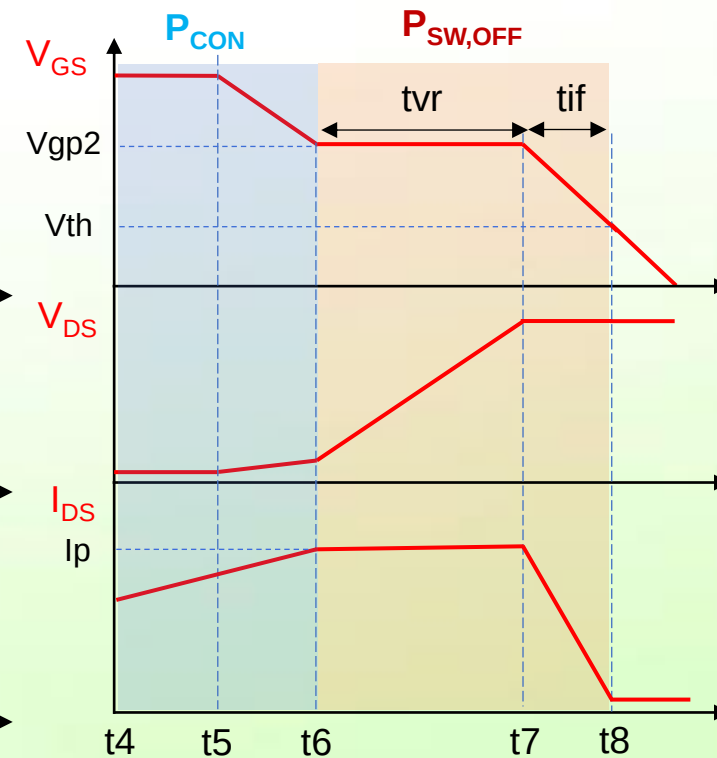
$$t_{vr} = R_G \frac{Q_{GD}}{V_{DS}} * \frac{V_{DS}}{V_{gp2}}$$

$$t_{if} = R_G C_{iss} * \ln \left(\frac{V_{gp2}}{V_{TH}} \right)$$

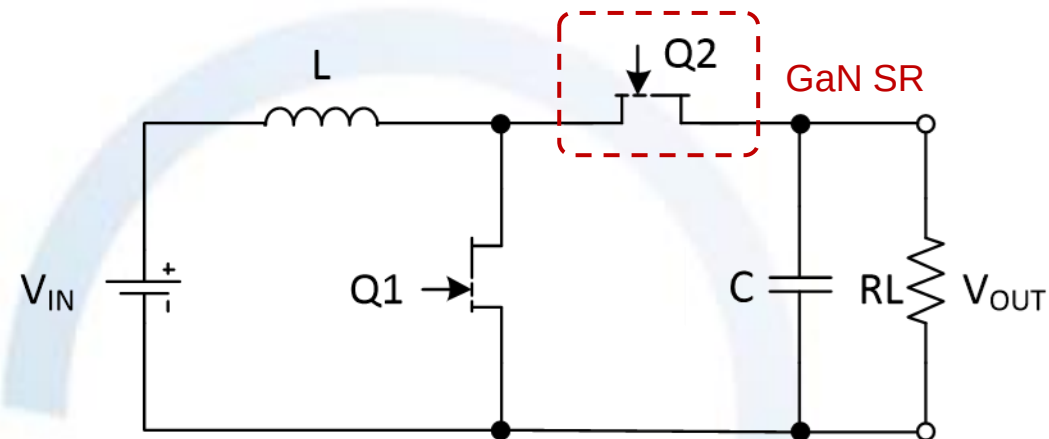
Turn-on transient



Turn-off transient



Loss breakdown of GaN FETs: SR conduction loss

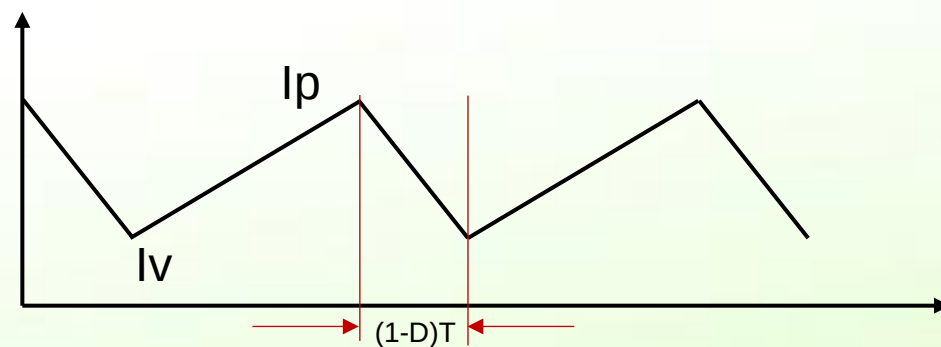


$$I_p = I_L + \frac{\Delta I_L}{2} \quad I_v = I_L - \frac{\Delta I_L}{2}$$

$$I_{rms, sr} = \sqrt{\frac{I_p^2 + I_v^2 + I_p I_v}{3}} * (1 - D)$$

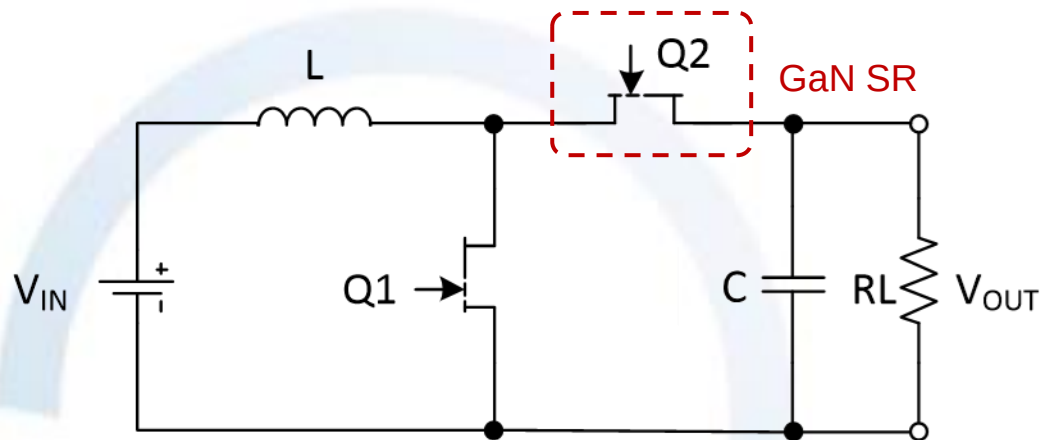
$$P_{cond, sr} = I_{rms, sr}^2 * R_{dson} = \frac{I_p^2 + I_v^2 + I_p * I_v}{3} * (1 - D) * R_{dson}$$

Similar to RMS current derivation of SW



$$I_{rms, sr} = \sqrt{\frac{I_p^2 + I_v^2 + I_p I_v}{3}} (1 - D)$$

Loss breakdown of GaN FETs: SR deadtime loss

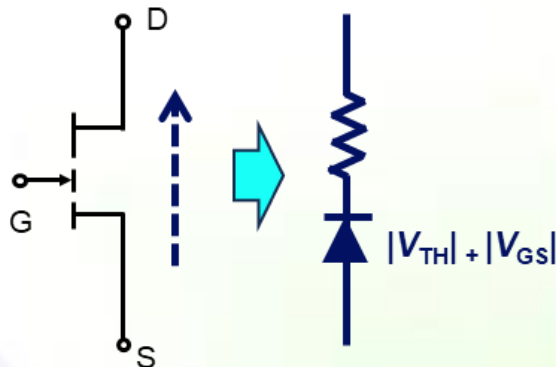


For turn-on and turn-off transient, the load current are different (I_p , I_v). If deadtime is a t_{dt}

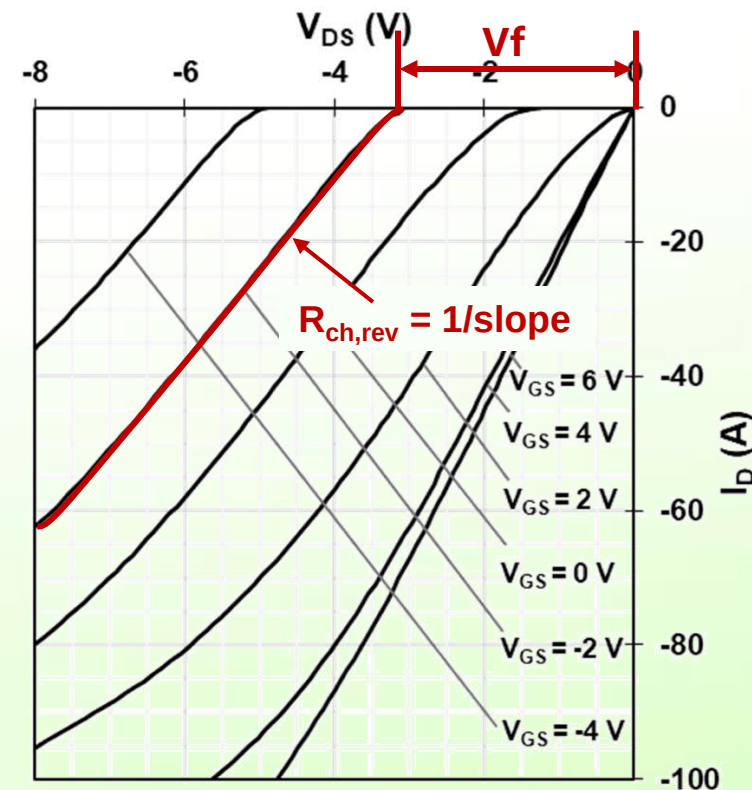
The deadtime loss is:

$$P_{dt} = (I_p * V_{SD_Ip} * t_{dt} + I_v * V_{SD_Iv} * t_{dt}) * fs$$

If $V_{GS} < V_{TH}$



- $V_f = |V_{TH}| + |V_{GS,OFF}|$
- $V_{SD} = V_f + I_L * R_{ch,rev}$

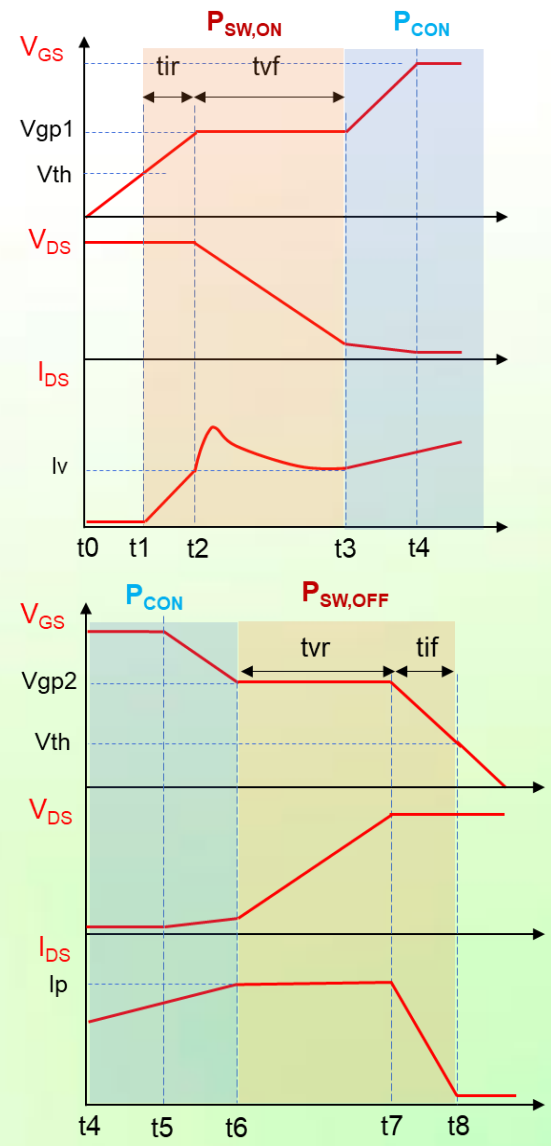


Boost converter loss summary



	Loss type	Equations
SW-Q1	Conduction loss	$P_{cond,sw} = \frac{I_p^2 + I_v^2 + I_p * I_v}{3} * D * Rdson$
	Switching loss	$P_{SW} \approx \frac{1}{2} * V_{IN} * I_v * t_{on} * f_s + \frac{1}{2} * V_{IN} * I_p * t_{off} * f_s$
	Capacitive loss	$P_{CAP} = Q_{OSS} * V_{bus} * f_s$
	Gate drive loss	$P_{GATE} = Q_G * V_{GS} * f_s$
SR-Q2	Conduction loss	$P_{cond,sr} = \frac{I_p^2 + I_v^2 + I_p * I_v}{3} * (1 - D) * Rdson$
	Deadtime loss	$P_{dt} = (I_p * V_{SD_Ip} * t_{dt} + I_v * V_{SD_Iv} * t_{dt}) * f_s$
	Gate drive loss	$P_{GATE} = (Q_G - Q_{GD}) * V_{GS} * f_s$

* Gate drive loss mainly happens in driver circuit, not in GaN FETs.



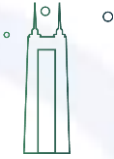
Powering the Dreams by CloudSemi GaN



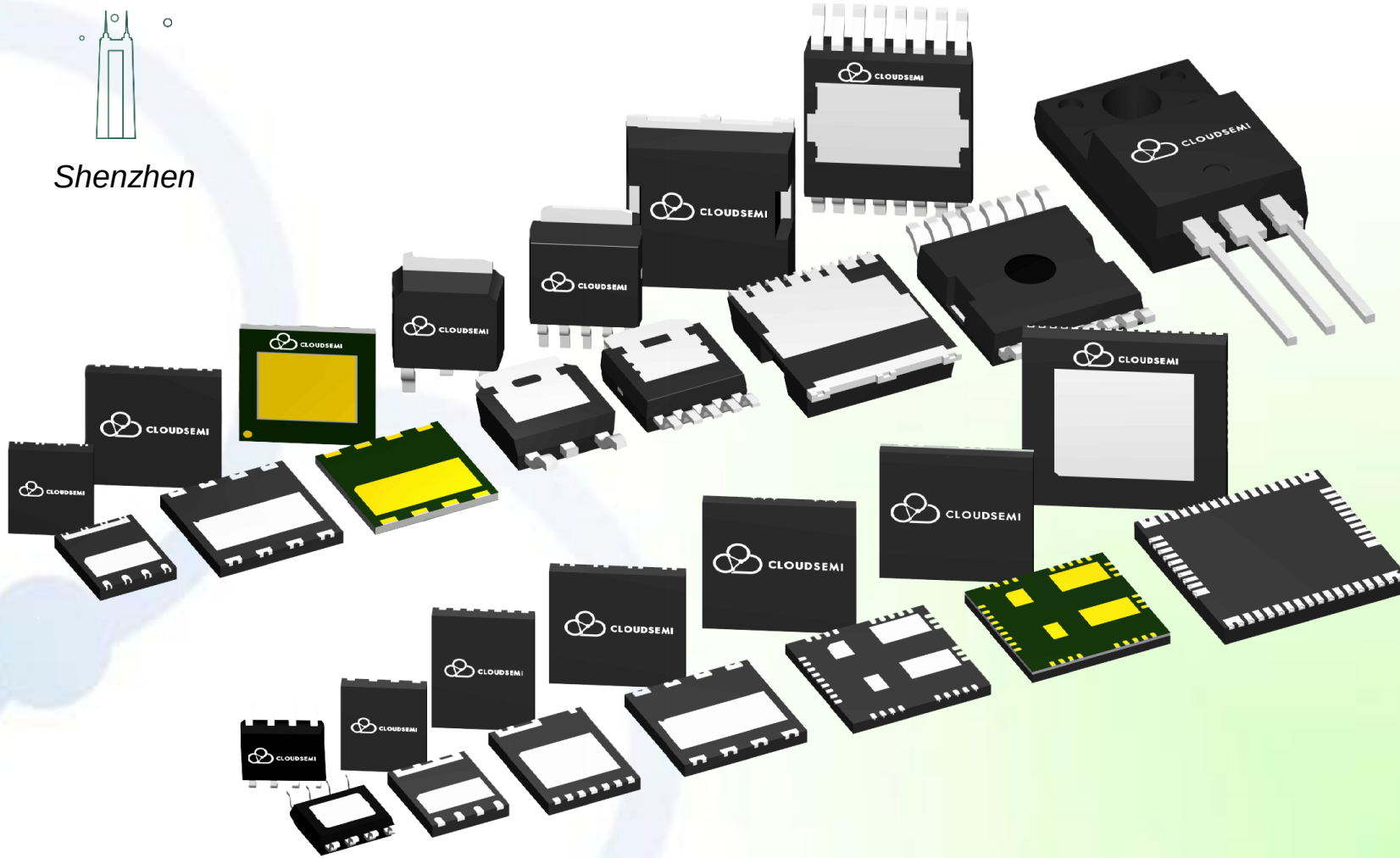
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Thank You!

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